

# Exploiting Structured Sparsity in Bayesian Experimental Design

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## Outline:

1. Compressive sensing under **structured** sparsity
2. **Adaptive** compressive sensing via **Bayesian experimental design**
3. **Approximate message passing (AMP)** for structured-sparse recovery
4. How to make AMP (and other algorithms like LASSO) adaptive
5. Empirical performance close to **oracle bounds**.

## Compressive Sensing:

- In compressive sensing, we aim to recover a signal vector  $\mathbf{u}$  from noisy **underdetermined** linear measurements

$$\mathbf{y} = \Phi \mathbf{u} + \mathbf{w} \in \mathbb{R}^M.$$

- Although the problem is underdetermined, accurate recovery maybe possible if  $\mathbf{u}$  can be **sparsely** represented in some dictionary  $\Psi$ , i.e.,

$$\mathbf{u} = \Psi \mathbf{x} \text{ for } K\text{-sparse } \mathbf{x} \in \mathbb{R}^N,$$

where  $\Psi$  is “incoherent” with  $\Phi$ .

- It is common to choose  $\Phi$  **randomly** and apply the **LASSO** algorithm to recover an estimate  $\hat{\mathbf{x}}$ , in which case one can guarantee  $\|\hat{\mathbf{x}} - \mathbf{x}\|_2^2 \leq C \|\mathbf{w}\|_2^2$ , for some constant  $C$ , with

$$M \geq \mathcal{O}(K \log(N/K)) \text{ measurements.}$$

## Structured Sparsity:

- Often the signal  $u$  has a representation  $x$  that is not simply sparse but rather **structured sparse**.

For examples,

- wavelet coefficients of natural images are **tree-sparse**, and
  - impulse responses of wideband wireless channels are **clustered-sparse**.
- In this case, similar reconstruction guarantees are possible with only

$$M \geq \mathcal{O}(K) \text{ measurements}$$

using structured-sparse recovery algorithms!

## Adaptive Compressive Sensing:

- In some applications, we can afford  $T > 1$  measurement rounds and **adapt** the measurement matrix  $\Phi_t$  for the  $t^{\text{th}}$  round based on the knowledge gained from previous rounds.
- In this case, the observation model changes to

$$\underbrace{\begin{bmatrix} \underline{y}_{t-1} \\ \underline{y}_t \end{bmatrix}}_{\underline{\mathbf{y}}_t} = \underbrace{\begin{bmatrix} \underline{\Phi}_{t-1} \\ \underline{\Phi}_t \end{bmatrix}}_{\underline{\Phi}_t} \mathbf{u} + \underbrace{\begin{bmatrix} \underline{\mathbf{w}}_{t-1} \\ \underline{\mathbf{w}}_t \end{bmatrix}}_{\underline{\mathbf{w}}_t \in \mathbb{R}^{M_t}} \begin{matrix} \in \mathbb{R}^{M_{t-1}} \\ \in \mathbb{R}^{M_t} \end{matrix},$$

where underbars are used to denote **cumulative** quantities.

*So, how is  $\Phi_t$  designed?*

- In Bayesian experimental design [DeGroot 62],  $\Phi_t$  is chosen to maximize the **expected information gain (EIG)**.

## Bayesian Experimental Design:

- The **information gain** is defined as the **KL divergence** between the **prior** and **posterior** distributions at measurement step  $t$ :

$$D(\mathbf{y}_t) \triangleq \int_{\mathbf{u}} q(\mathbf{u} | \mathbf{y}_t) \log \frac{q(\mathbf{u} | \mathbf{y}_t)}{q(\mathbf{u})},$$

where

$q(\mathbf{u}) \triangleq p(\mathbf{u} | \underline{\mathbf{y}}_{t-1})$  is the step- $t$  **prior**, and

$q(\mathbf{u} | \mathbf{y}_t) \triangleq p(\mathbf{u} | \underline{\mathbf{y}}_{t-1}, \mathbf{y}_t)$  is the step- $t$  **posterior**.

- Since  $\mathbf{y}_t$  is not yet known, we consider **expected** information gain:

$$\begin{aligned} \text{EIG}_t &\triangleq \mathbb{E}\{D(\mathbf{y}_t) | \underline{\mathbf{y}}_{t-1}\} = \int_{\mathbf{y}_t} \underbrace{p(\mathbf{y}_t | \underline{\mathbf{y}}_{t-1})}_{\triangleq q(\mathbf{y}_t)} \int_{\mathbf{u}} q(\mathbf{u} | \mathbf{y}_t) \log \frac{q(\mathbf{u} | \mathbf{y}_t)}{q(\mathbf{u})} \\ &= \int_{\mathbf{y}_t} \int_{\mathbf{u}} q(\mathbf{u}, \mathbf{y}_t) \log \frac{q(\mathbf{u}, \mathbf{y}_t)}{q(\mathbf{u})q(\mathbf{y}_t)} = \text{I}(\mathbf{u}; \mathbf{y}_t), \end{aligned}$$

i.e., the **mutual information** between  $\mathbf{u} \sim q(\mathbf{u})$  and  $\mathbf{y}_t \sim q(\mathbf{y}_t)$ .

## Gaussian Experimental Design:

- Evaluating the expected information gain is often **difficult**.
- However, when all distributions are **Gaussian**, it becomes easy.

For example, if

$$\text{noise: } \mathbf{w} \sim \mathcal{N}(\mathbf{0}, v_w \mathbf{I})$$

$$\text{step-}t \text{ signal prior: } \mathbf{u} | \mathbf{y}_{t-1} \sim \mathcal{N}(\boldsymbol{\mu}_u, \boldsymbol{\Sigma}_u)$$

then it is straightforward to show that

$$\text{EIG}_t = \frac{1}{2} \log \left| \frac{1}{v_w} \boldsymbol{\Phi}_t \boldsymbol{\Sigma}_u \boldsymbol{\Phi}_t^T + \mathbf{I} \right|.$$

- Of course, in compressive sensing, the signal priors are **non-Gaussian** and thus the above could only be used after approximations are made.

## Gaussian design of $\Phi_t$ :

What is the EIG-maximizing  $\Phi_t$  subject to the energy constraint  $\|\Phi_t\|_F^2 \leq \mathcal{E}$ ?

- Previous works [Seeger 08, Ji/Xu/Carin 08] studied the case of one **scalar** measurement per step (i.e.,  $M_t = 1$ ).

In this case,  $\Phi_t$  is a row vector and so  $\text{EIG}_t = \frac{1}{2} \log \left| \frac{1}{v_w} \Phi_t \Sigma_u \Phi_t^\top + \mathbf{I} \right|$  is maximized by the **dominant eigenvector** of  $\Sigma_u$ .

- In practice, though, we may want  $M_t \gg 1$  measurements per step. For this case, we show that the EIG is maximized by **waterfilling**:

**Lemma 1** Say that  $(\lambda_m, \mathbf{v}_m)_{m=1}^{M_t}$  are the  $M_t$  dominant (eigenvalue, eigenvector) pairs of  $\Sigma_u$ . Then for  $\{E_m\}_{m=1}^{M_t}$  and “water level”  $L$  satisfying

$$E_m = \max \{L - v_w / \lambda_m, 0\} \quad \forall m \in \{1, \dots, M_t\}$$

$$\sum_{m=1}^{M_t} E_m = \mathcal{E},$$

the  $m^{\text{th}}$  row of the EIG-maximizing  $\Phi_t$  equals  $\sqrt{E_m} \mathbf{v}_m$ .



## Leveraging Gaussian design for Adaptive CS:

- In CS, the step- $t$  prior (i.e., step- $(t-1)$  posterior)  $p(\mathbf{u} | \underline{\mathbf{y}}_{t-1})$  is non-Gaussian, and so a **Gaussian posterior approximation** must be made.
- Previous works have tackled this using a **Gaussian prior approximation**:
  - Say  $p(\mathbf{x} | \underline{\mathbf{y}}_{t-2}) \approx \prod_{n=1}^N \mathcal{N}(x_n; 0, \alpha_n^{-1})$  with “precision”  $\alpha_n$ .
  - Then  $p(\mathbf{x} | \underline{\mathbf{y}}_{t-1}) \approx \mathcal{N}(\mathbf{x}; \boldsymbol{\mu}_x, \boldsymbol{\Sigma}_x)$  with

$$\begin{aligned}\boldsymbol{\Sigma}_x &\triangleq \left( \frac{1}{v_w} \underline{\mathbf{A}}_{t-1}^\top \underline{\mathbf{A}}_{t-1} + \mathcal{D}(\boldsymbol{\alpha}) \right)^{-1} \\ \boldsymbol{\mu}_x &\triangleq \frac{1}{v_w} \boldsymbol{\Sigma}_x \underline{\mathbf{A}}_{t-1}^\top \underline{\mathbf{y}}_{t-1} \\ \underline{\mathbf{A}}_{t-1} &\triangleq \underline{\boldsymbol{\Phi}}_{t-1} \boldsymbol{\Psi}\end{aligned}$$

and so  $p(\mathbf{u} | \underline{\mathbf{y}}_{t-1}) \approx \mathcal{N}(\mathbf{u}; \boldsymbol{\mu}_u, \boldsymbol{\Sigma}_u)$  with  $\boldsymbol{\mu}_u = \boldsymbol{\Psi} \boldsymbol{\mu}_x$  and  $\boldsymbol{\Sigma}_u = \boldsymbol{\Psi} \boldsymbol{\Sigma}_x \boldsymbol{\Psi}^\top$ .

- To estimate  $\boldsymbol{\alpha}$ , [Ji/Xu/Carin 08] used Tipping’s RVM (“Bayesian CS”).
- Other works used different Gaussian posterior approximations:
  - [Seeger 08] assumed Laplacian  $\mathbf{x}$  and expectation propagation, and
  - [Seeger/Nickisch 11] used variational methods.

## Approximate Message Passing:

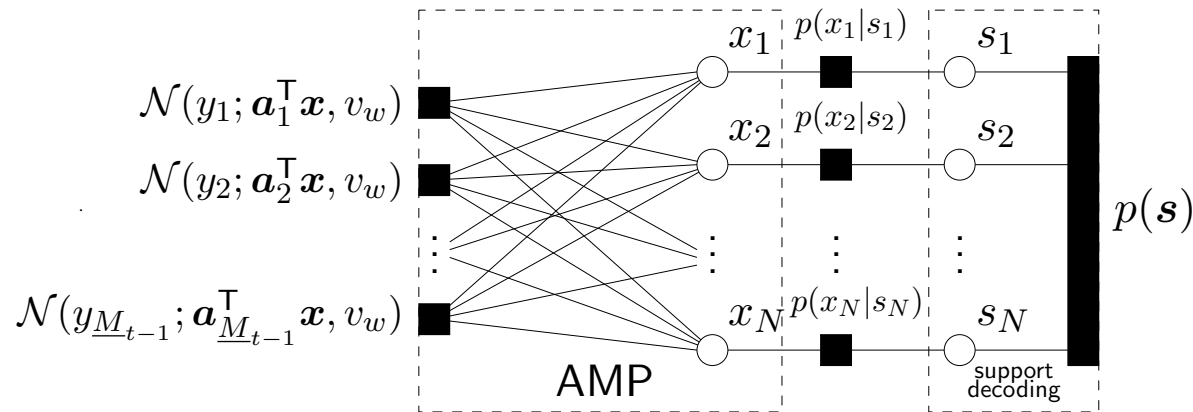
- Efficient sparse reconstruction algorithms have been constructed using loopy belief propagation with carefully constructed message approximations:
  - The **LASSO AMP** [Donoho/Maleki/Montanari 09] assumes i.i.d Laplacian signal, Gaussian noise, and i.i.d constructed  $\mathbf{A}$ .
  - The **Bayesian AMP** [Donoho/Maleki/Montanari 10] accepts generic signal priors, Gaussian noise, and i.i.d constructed  $\mathbf{A}$ .
  - The **generalized AMP** [Rangan 10] accepts generic signal and noise priors and arbitrary  $\mathbf{A}$ . (*We need this one!*)
- These AMP algorithms are **very fast iterative thresholding** algorithms. Their complexity is dominated by one application of  $\mathbf{A}$  and  $\mathbf{A}^T$  per iteration, and  $\lesssim 50$  iterations (for any  $M$  and  $N$ ) ... many fewer than FISTA.

## Turbo-AMP for Structured Sparsity:

- AMP has been extended to generic **structured-sparse** reconstruction using an approach inspired by **turbo** equalization and decoding.
- For this, the prior pdf is chosen as  $p(\mathbf{x}) = p(\mathbf{s}) \prod_{n=1}^N p(x_n | s_n)$  with a generic support prior  $p(\mathbf{s})$  and Bernoulli-Gaussian amplitudes:

$$p(x_n | s_n) = s_n \mathcal{N}(x_n; 0, v_x) + (1 - s_n) \delta(x_n), \quad s_n \in \{0, 1\}.$$

In this case, the factor graph becomes



and we pass **extrinsic likelihoods** on  $\{s_n\}$  back and forth between the two soft-input/soft-output “decoders” [Schniter 10].

## Turbo-AMP for Adaptive CS:

- To leverage Gaussian experiment design, we propose a variation on the **Gaussian prior approximation** used in [Ji/Xu/Carin 08]:

$$p(\mathbf{x} | \underline{\mathbf{y}}_{t-2}) \approx \prod_{n=1}^N \mathcal{N}(x_n; 0, \alpha_n^{-1})$$

- Instead of using the RVM to ML-estimate  $\{\alpha_n\}$ , we use **AMP's marginal posteriors**

$$p(x_n | \underline{\mathbf{y}}_{t-1}) \approx \mathcal{N}(x_n; \hat{x}_n, \nu_n) \quad \text{and} \quad \Pr\{s_n=1 | \underline{\mathbf{y}}_{t-1}\} \approx \lambda_n.$$

In particular, we propose several **surrogates** for the inverse precisions  $\alpha_n^{-1}$ :

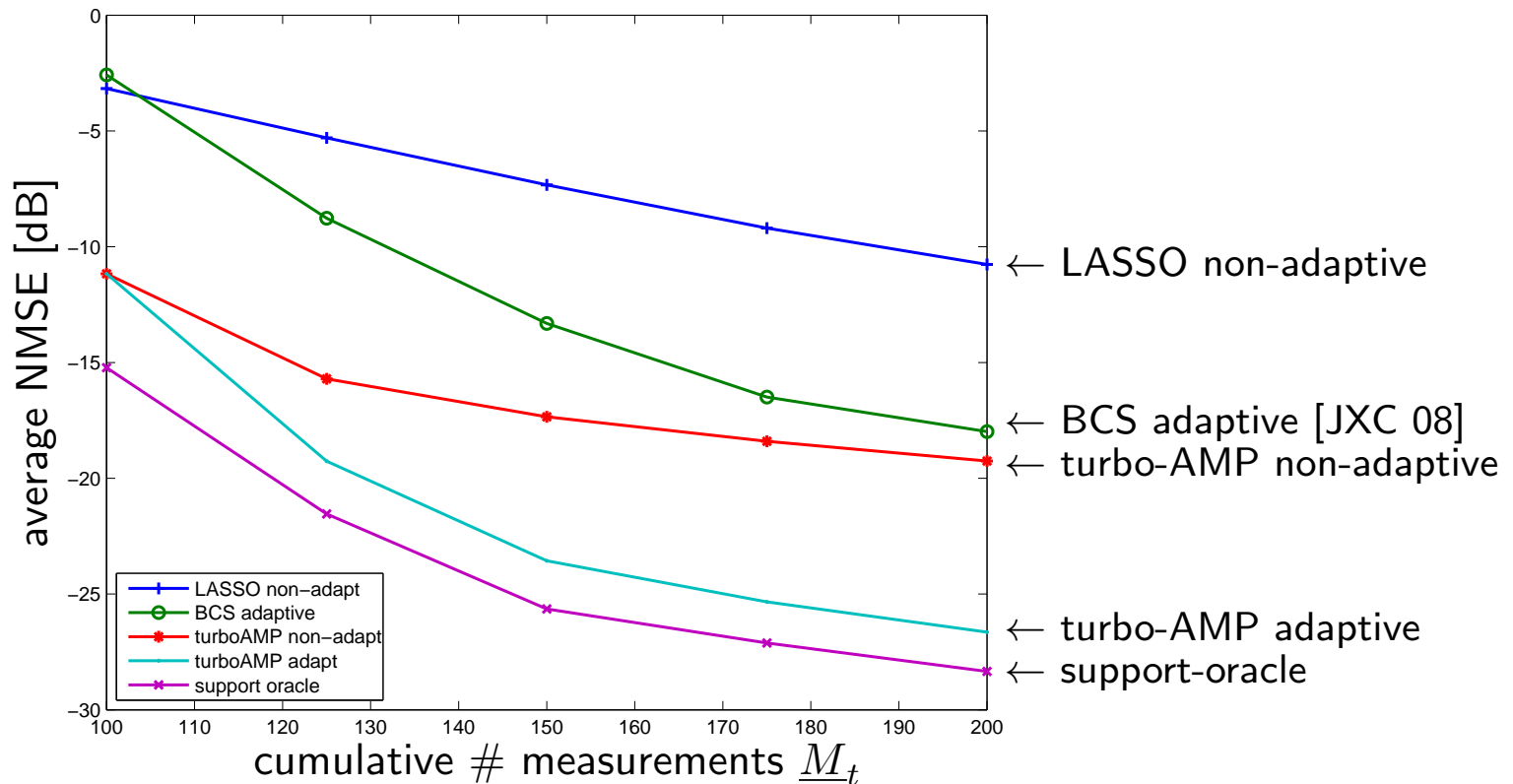
- “Variance”:  $\hat{\alpha}_n^{-1} = \nu_n$ .
- “Mean”:  $\hat{\alpha}_n^{-1} = |\hat{x}_n|^2$  ... only point estimates ( $\rightsquigarrow$  **adaptive Lasso!**)
- “Energy”:  $\hat{\alpha}_n^{-1} = |\hat{x}_n|^2 + \nu_n$
- “Support”:  $\hat{\alpha}_n^{-1} = \lambda_n v_x$ ,

## Empirical Study:

We now present empirical evidence showing that the proposed **adaptive turbo-AMP** performs very close to **oracle bounds**.

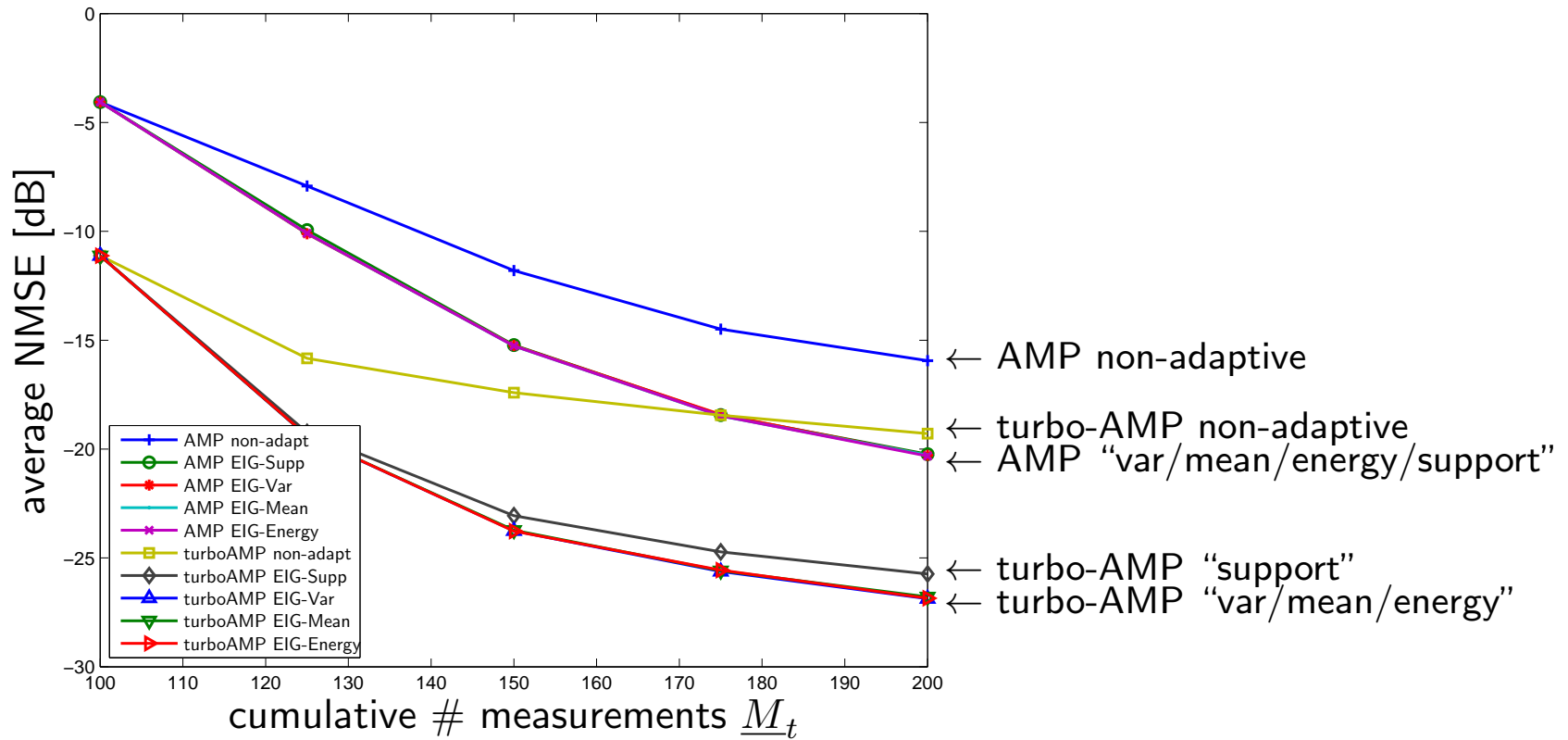
- Clustered-sparse Bernoulli-Gaussian signal:
  - length  $N = 500$ ,
  - sparsity  $K = 50$ ,
  - average cluster-size = 11.
- Canonical sparsifying dictionary  $\Psi = \mathbf{I}$  (i.e.,  $\mathbf{u} = \mathbf{x}$ ).
- AWGN yielding average SNR = 15dB.
- $T=5$  measurement steps, with  $M_0=100$  i.i.d- $\mathcal{N}$ , then subsequently  $M_t=50$ .
- We report NMSE  $\|\hat{\mathbf{x}} - \mathbf{x}\|_2^2 / \|\mathbf{x}\|_2^2$  averaged over 500 realizations.
- We compare to the **support oracle**, for which signal is Gaussian, and so both EIG-maximizing  $\Phi_t$  and MSE-minimizing  $\hat{\mathbf{x}}$  can be computed in closed form.

## NMSE versus cumulative measurements $\underline{M}_t$ :



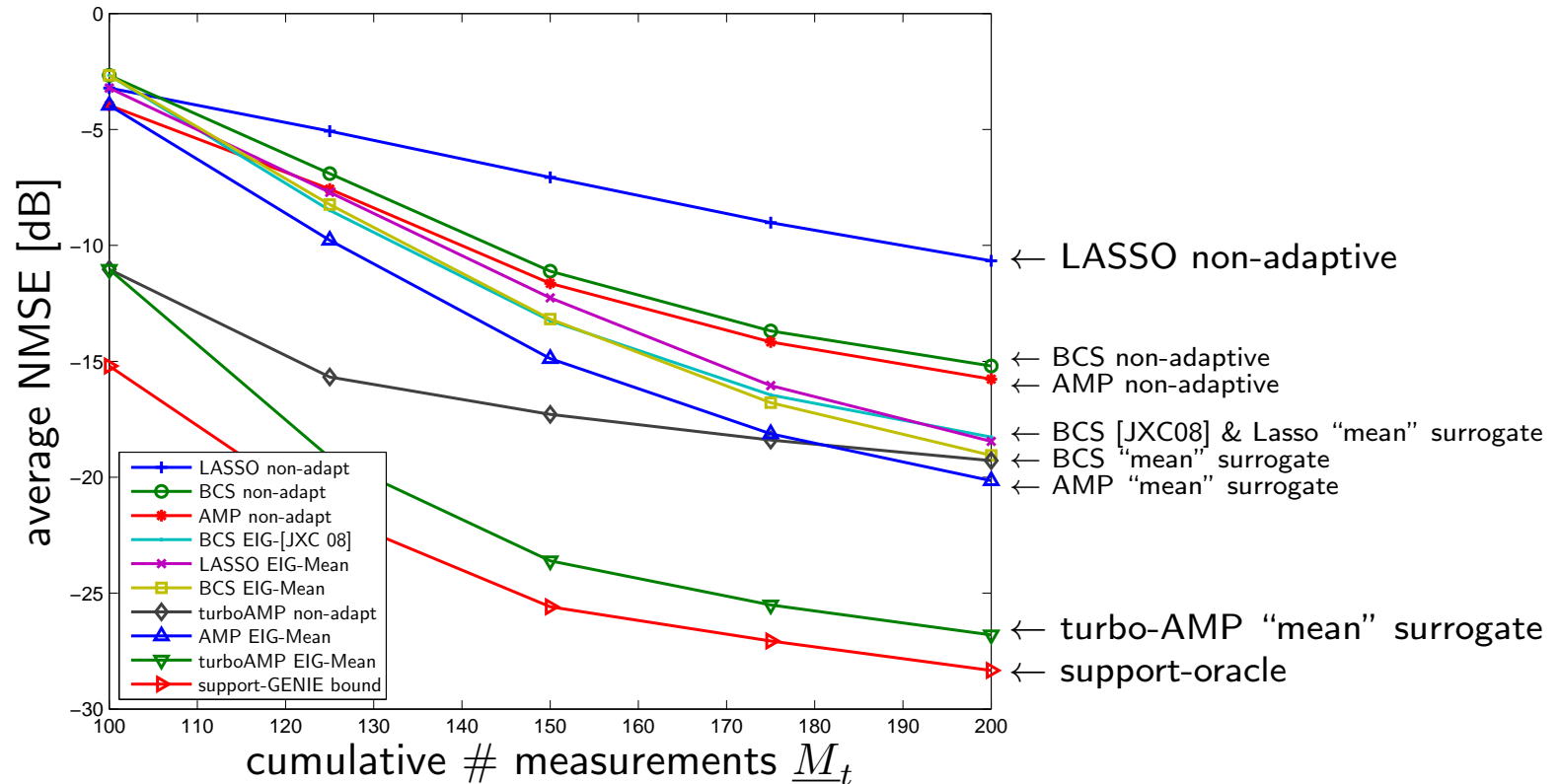
- Performances gain from structured sparsity, adaptivity, and the combination.
- **Adaptive turbo-AMP performs 1.5 dB from the support-oracle bound!**

## Effect of surrogate choice in Gaussian prior approximation:



Relatively **insensitive** to the Gaussian-prior-approximation used in  $\Phi_t$  design.

## Using the “mean” surrogate to create new algorithms:



- Adaptation using our “mean” surrogate yields an **adaptive LASSO**.
- Adaptation using our “mean” surrogate **improves** BCS over [JXC 08].



## Summary and ongoing work:

- Main focus:
  - Merging **Bayesian experim. design** with **structured-sparse** recovery.
- Contributions:
  - **Waterfilling** solves Gaussian experimental design for  $M_t > 1$  meas/step.
  - Novel adaptation heuristics leading to **adaptive LASSO**, etc.
  - An **adaptive turbo-AMP** empirically performing near **oracle bounds**.
- Ongoing work:
  - Optimal design of **initial**  $\Phi_0$ .
  - Theoretical analysis using AMP's **state evolution**.
  - Extension to **pre**-measurement noise model  $\mathbf{y} = \Phi(\Psi\mathbf{x} + \mathbf{v}) + \mathbf{w}$ .
  - Adaptation under **constrained**  $\Phi$  (e.g., Toeplitz).
  - Development/analysis of **simplified** schemes (no eigendecomposition).

*Thanks!*