Decision Feedback Equalization:

- A technique with performance close to MLWD but complexity close to linear equalization.
- The DFEs derived here are in "standard" form, and so my development differs from that of Fitz!
- Assume front-end matched-filtering that yields $\{\tilde{Q}^{(k)}\}$, giving the effective discrete-time channel G(z):

$$E_b D_z^{(k)} \longrightarrow G(z) \longrightarrow \tilde{Q}^{(k)}$$

$$\tilde{N}^{(k)}$$

where we recall that $S_{\tilde{N}}(e^{j2\pi f}) = E_b N_o G(e^{j2\pi f})$.

1

Phil Schniter OSU ECE-809

Minimum-Phase Spectral Factorization:

Any rational power spectrum S_x can be factored as follows:

$$S_x(z) = \gamma_x F_x^+(z) F_x^-(z)$$

$$F_x^+(z) = 1 + \sum_{l=1}^{\infty} f_x[l] z^{-l} \text{ (monic, causal, min-phase)}$$

$$F_x^-(z) = 1 + \sum_{l=1}^{\infty} f_x^*[l] z^l \text{ (monic, anti-causal, max-phase)}$$

$$\gamma_x = \exp\left[\int_{-0.5}^{0.5} \ln S_x(e^{j2\pi f}) df\right] = \langle S_x \rangle_{\mathsf{G}}$$

Notes:

- $F_x^-(z) = [F_x^+(1/z^*)]^*$, $F_x^-(e^{j2\pi f}) = [F_x^+(e^{j2\pi f})]^*$.
- $\bullet \ [F_x^+(z)]^{-1}$ is causal, $[F_x^-(z)]^{-1}$ is anti-causal.

The Whitened Matched Filter Front-end:

Idea: Whiten the noise $\{\tilde{N}^{(k)}\}$ in the discrete time model.

$$S_{\tilde{N}}(e^{j2\pi f}) = E_b N_o G(e^{j2\pi f})$$

$$= E_b N_o \gamma_G F_G^+(e^{j2\pi f}) F_G^-(e^{j2\pi f})$$

$$E_b D_z^{(k)} \rightarrow G(z) \rightarrow \tilde{Q}^{(k)} \qquad [\gamma_G F_G^-(z)]^{-1} \rightarrow \bar{Q}^{(k)}$$

$$\tilde{N}^{(k)}$$

Note the causal monic channel and uncorrelated noise:

$$E_b D_z^{(k)} \longrightarrow \boxed{F_G^+(z)} \longrightarrow \bar{Q}^{(k)}$$

$$\tilde{W}^{(k)}$$

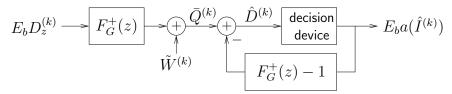
$$S_{\tilde{W}}(e^{j2\pi f}) = E_b N_o \gamma_G^{-1}$$

3

Phil Schniter OSU ECE-809

The ZF Decision Feedback Equalizer:

Cancel the *post-cursor* ISI after the WMF front-end:



Notice that the feedback filter is *strictly causal*. Without error propagation, $\hat{D}^{(k)} = E_b D_z^{(k)} + \tilde{W}^{(k)}$, so that

$$PBE = \frac{1}{2}\operatorname{erfc}\left(\sqrt{\frac{E_b\gamma_G}{N_o}}\right)$$

Compare to ZF-LE:

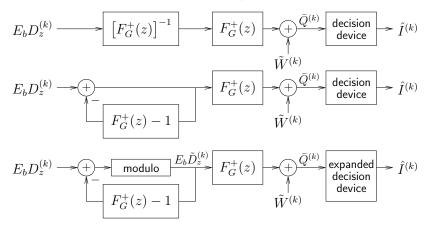
$$\frac{\gamma_G}{\langle G^{-1}\rangle_{\mathsf{A}}^{-1}} = \frac{\langle G\rangle_{\mathsf{G}}}{\langle G^{-1}\rangle_{\mathsf{A}}^{-1}} = \frac{\langle G^{-1}\rangle_{\mathsf{G}}^{-1}}{\langle G^{-1}\rangle_{\mathsf{A}}^{-1}} = \frac{\langle G^{-1}\rangle_{\mathsf{A}}}{\langle G^{-1}\rangle_{\mathsf{G}}} \ge 1.$$

4

Tomlinson-Harashima Precoding:

• If channel known at transmitter, can pre-equalize (i.e., "pre-code") to eliminate ISI at receiver:

• To avoid increase in transmitted power, use modulo transmission and periodically expand the decision device:



OSU ECE-809

5

Phil Schniter

The MMSE Decision Feedback Equalizer:

Recall the linear MMSE equalizer: $W(z) = \left(G(z) + \frac{N_o}{E_b}\right)^{-1}$

$$E_b D_z^{(k)} \longrightarrow \boxed{G(z)} \longrightarrow \hat{\tilde{Q}}^{(k)} \boxed{W(z)} \stackrel{\hat{D}^{(k)}}{\longrightarrow} \boxed{\text{decision device}} \rightarrow \hat{I}^{(k)}$$

Say $\hat{D}^{(k)} = E_b D_z^{(k)} + E^{(k)}$. Detection performance will increase if we reduce σ_E^2 . Can maximally reduce σ_E^2 by whitening $\{E^{(k)}\}$ using optimal linear prediction:

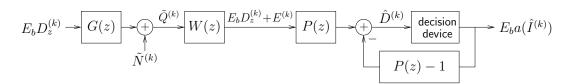
$$E^{(k)} \longrightarrow P(z) = (F_E^+(z))^{-1} \longrightarrow \tilde{E}^{(k)}$$

$$S_E(e^{j2\pi f}) = \gamma_E F_E^+(e^{j2\pi f}) F_E^-(e^{j2\pi f}), \quad \sigma_E^2 = \gamma_E \sum_{l=0}^{\infty} |f[l]|^2$$

 $S_{\tilde{E}}(e^{j2\pi f}) = \gamma_E, \quad \sigma_{\tilde{E}}^2 = \gamma_E$

6

Cleverly, we can whiten $\{E^{(k)}\}$ without affecting $E_b D_z^{(k)}$!



Assuming perfect decisions,

$$\hat{D}^{(k)} = P(z) \left[E_b D_z^{(k)} + E^{(k)} \right] - \left[P(z) - 1 \right] E_b D_z^{(k)}
= E_b D_z^{(k)} + P(z) E^{(k)}$$

Note that the feedback filter is strictly causal:

P(z) is causal and monic because $P(z) = (F_E^+(z))^{-1}$.

So what is $F_E^+(z)$?

7

Phil Schniter OSU ECE-809

Recall from MMSE equalization that

$$W(z) = \left(G(z) + \frac{N_o}{E_b}\right)^{-1} \Leftrightarrow 1 = G(z)W(z) + \frac{N_o}{E_b}W(z)$$

which is useful in simplifying

$$E^{(k)} = (G(z)W(z) - 1)E_b D_z^{(k)} + W(z)\tilde{N}^{(k)}$$

= $-W(z)N_o D_z^{(k)} + W(z)\tilde{N}^{(k)}$.

Assuming zero-mean white $\{D_z^{(k)}\}$, independent of $\tilde{N}^{(k)}$,

$$S_{E}(e^{j2\pi f}) = W^{2}(e^{j2\pi f}) \left(N_{o}^{2} + E_{b}N_{o}G(e^{j2\pi f})\right)$$

$$= E_{b}N_{o}W(e^{j2\pi f})$$

$$= E_{b}N_{o}\gamma_{W}F_{W}^{+}(e^{j2\pi f})F_{W}^{-}(e^{j2\pi f})$$

$$\Rightarrow F_{E}^{+}(z) = F_{W}^{+}(z)$$

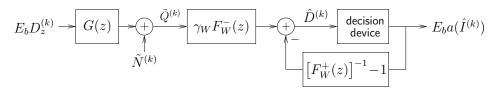
8

The MMSE-DFE forward filter equals

$$W(z)P(z) = W(z)(F_W^+(z))^{-1} = \gamma_W F_W^-(z)$$

and so is referred to as a "precursor equalizer."

Summary of MMSE-DFE:



9

Phil Schniter OSU ECE-809

It is interesting to note that the MMSE estimate is biased:

$$\hat{D}^{(k)} = E_b D_z^{(k)} + P(z) E^{(k)}
= E_b D_z^{(k)} - P(z) W(z) N_o D_z^{(k)} + P(z) W(z) \tilde{N}^{(k)}
= \left(1 - \frac{N_o}{E_b} \gamma_W F_W^-(z)\right) E_b D_z^{(k)} + \gamma_W F_W^-(z) \tilde{N}^{(k)}
= \left(1 - \frac{N_o}{E_b} \gamma_W\right) E_b D_z^{(k)} - \frac{N_o}{E_b} \gamma_W \sum_{l=1}^{\infty} f_W^*[l] D_z^{(k+l)}
+ \gamma_W F_W^-(z) \tilde{N}^{(k)}$$

Bias removal will decrease PBE (and increase MSE).

→ unbiased MMSE-DFE

Unbiasing can be incorporated into the "decision device."