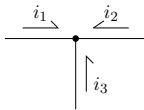


**Ohm's law:**

$$v = iR, \quad i = vG$$

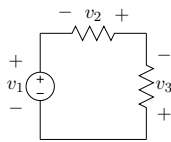
**Kirchoff's law's:**

**KCL**



$$i_1 + i_2 + i_3 = 0$$

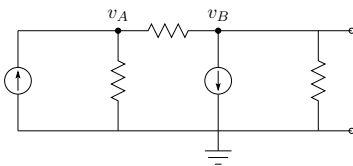
**KVL**



$$v_1 + v_2 + v_3 = 0$$

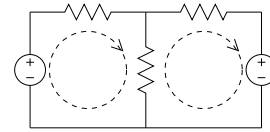
**Node voltage analysis:**

1. KCL at each node
2. substitute  $i = vG$  at each resistor
3. solve system of equations for node voltages



**Mesh current analysis:**

1. KVL around each loop
2. substitute  $v = iR$  at each resistor
3. solve system of equations for mesh currents

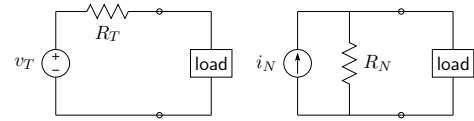


**Superposition:**

1. solve circuit assuming only one voltage/current source is active
  2. add voltages/currents from each individual solution
- ... a consequence of our focus on *linear* circuits.

**Thevenin/Norton equivalents:**

Possible to reduce linear sources to either...



$v_T$  = open circuit voltage

$i_N$  = short circuit current

$$R_T = R_N = \frac{v_T}{i_N}$$

**Capacitance:**

$$i_c(t) = C \frac{dv_c(t)}{dt}$$

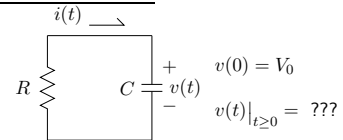
1. constant voltage across  $\Leftrightarrow$  no current thru
2. no sudden voltage changes (would require infinite current!)
3. can store as well as release energy

**Inductance:**

$$v_l(t) = L \frac{di_l(t)}{dt}$$

1. fixed current thru  $\Leftrightarrow$  no voltage across
2. no sudden current changes (would require infinite voltage!)
3. can store as well as release energy

**RC response to initial conditions:**



KVL yields (for  $t \geq 0$ )

$$0 = v(t) + i(t)R$$

$$= v(t) + RC \frac{dv(t)}{dt} \quad \dots \text{a differential equation}$$

Trying  $v(t) = Ke^{st}, t \geq 0$

$$0 = Ke^{st} + RCKe^{st}s = K(1 + RCs)e^{st}$$

$$\Rightarrow s = -1/RC$$

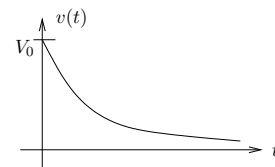
$$\Rightarrow v(t) = Ke^{-t/RC}$$

Applying initial condition:

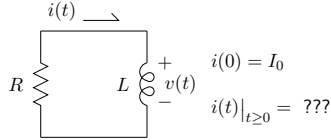
$$V_0 = v(0) = Ke^0 = K$$

$$\Rightarrow v(t) = V_0 e^{-t/RC}, t \geq 0$$

Graphically:



**RL response to initial conditions:**



KCL yields (for  $t \geq 0$ )

$$0 = i(t) + v(t)G$$

$$= i(t) + GL \frac{di(t)}{dt} \dots \text{a differential equation}$$

Trying  $i(t) = Ke^{st}, t \geq 0$

$$0 = Ke^{st} + GLKse^{st} = K(1 + GLs)e^{st}$$

$$\Rightarrow s = -1/GL$$

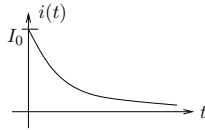
$$\Rightarrow i(t) = Ke^{-t/GL}$$

Applying initial condition:

$$I_0 = i(0) = Ke^0 = K$$

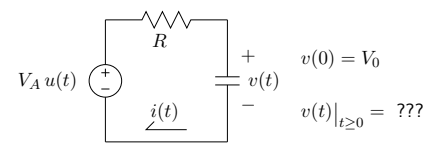
$$\Rightarrow i(t) = I_0 e^{-t/GL}, t \geq 0$$

Graphically:



Dual of the RC case!

**RC response to step function & initial condition:**



Main idea:  $v(t) = \underbrace{v_N(t)}_{\text{natural}} + \underbrace{v_F(t)}_{\text{forced}}$  (superposition!)

Natural response:

$$\text{diff eq: } 0 = RC \frac{dv_N(t)}{dt} + v_N(t), t \geq 0$$

$$\text{postulate } v_N(t) = K_1 e^{st}, t \geq 0$$

$$\Rightarrow v_N(t) = K_1 e^{-t/RC}, t \geq 0$$

Forced response:

$$\text{diff eq: } V_A = RC \frac{dv_F(t)}{dt} + v_F(t), t \geq 0$$

$$\text{postulate } v_F(t) = K_2, t \geq 0$$

$$\Rightarrow v_F(t) = V_A, t \geq 0$$

Together

$$v(t) = V_A + K_1 e^{-t/RC}, t \geq 0$$

Applying initial condition:

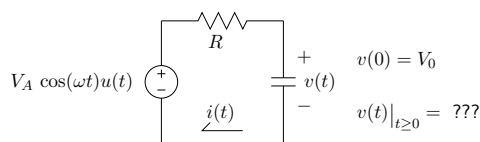
$$v(0) = V_0 \Rightarrow K_1 = V_0 - V_A$$

Thus

$$v(t) = V_A + (V_0 - V_A)e^{-t/RC}, t \geq 0$$

What about an RL circuit?

**RC response to causal sinusoid with initial condition:**



Again, we use  $v(t) = \underbrace{v_N(t)}_{\text{natural}} + \underbrace{v_F(t)}_{\text{forced}}$

Natural response:

$$\text{diff eq: } 0 = RC \frac{dv_N(t)}{dt} + v_N(t), t \geq 0$$

$$\Rightarrow v_N(t) = K_1 e^{-t/RC}, t \geq 0$$

Forced response:

$$\text{diff eq: } V_A \cos(\omega t) = RC \frac{dv_F(t)}{dt} + v_F(t), t \geq 0$$

$$\text{postulate } v_F(t) = K_2 \cos(\omega t + \phi), t \geq 0$$

$$= a \cos(\omega t) + b \sin(\omega t)$$

$$\text{solving diff eq gives } a = \frac{V_A}{1 + (\omega RC)^2}, b = \frac{\omega RC V_A}{1 + (\omega RC)^2}$$

Together

$$v(t) = K_1 e^{-t/RC} + a \cos(\omega t) + b \sin(\omega t), t \geq 0$$

Applying initial condition:

$$v(0) = V_0 \Rightarrow K_1 = V_0 - a$$

Thus, for  $t \geq 0$ ,

$$v(t) = \left( V_0 - \frac{V_A}{1 + (\omega RC)^2} \right) e^{-t/RC} + \frac{V_A}{1 + (\omega RC)^2} (\cos(\omega t) + \omega RC \sin(\omega t))$$

**Sinusoidal steady-state response:**

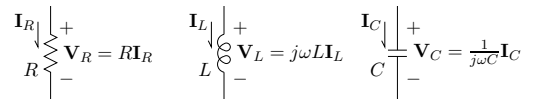
In an RLC circuit sinusoidally driven at frequency  $\omega$ , all current and voltage waveforms are sinusoidal with frequency  $\omega$  and thus completely described by their *magnitudes and phases*.

$\Rightarrow$  phasors

$$V_0 \cos(\omega t + \phi) = \text{Re}\{V_0 e^{j(\omega t + \phi)}\} = \text{Re}\{\underbrace{V_0 e^{j\phi}}_{\text{phasor } \mathbf{V}} \cdot e^{j\omega t}\}$$

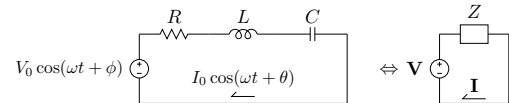
Passive circuit elements can be described in terms of their effect on current and voltage phasors:

$\Rightarrow$  impedance  $Z$ , where  $\mathbf{V} = \mathbf{Z}\mathbf{I}$



$$Z_R = R \quad Z_L = j\omega L \quad Z_C = \frac{1}{j\omega C}$$

Leads to simple (steady state sinusoidal) circuit analysis...



$$Z = R + (j\omega C)^{-1} + j\omega L$$

$$\mathbf{V} = V_0 e^{j\phi}$$

$$\Rightarrow \mathbf{I} = I_0 e^{j\theta} = \mathbf{V}/Z$$

... much easier than diff eq method!